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# A novel approach of order diminution using time moment concept with Routh array and salp swarm algorithm 

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#### Abstract

In control engineering, there may be two systems that have the same input-output characteristic with different degrees of complexity. This concept leads to order diminution(OD) of a large scale system. In this article, the authors propose a new hybrid order diminution technique based on the time moment matching method with the Routh array concept and a recently developed fast and accurate salp swarm optimization (SSO) technique. The proposed method combines the advantages of both the classical method of OD and the optimization technique. The unknown coefficient of the divisor of the reduced system is obtained by exploring the time moment matching methodology with the Routh array concept, whereas the unknown coefficients of dividend polynomial are obtained by the SSO technique. The time moment matching with the Routh array ensures the nature of the system in terms of stability, and better search capability of the SSO technique reduces the error between the original and the diminished system. The proposed technique is tested on different benchmark problems, including a time-delay system. An intensive comparative study in terms of different errors, time, and frequency domain provides better performance of the proposed method compared to the existing techniques. A good match to the parameters of transient specifications indicates the success of this proposed technique. Comparatively, the matching of rise time, settling time, and maximum overshoots are $99.6076 \%, 99.1611 \%$, and $100 \%$, respectively.


Key words: Error minimization, model order reduction, order diminution, Routh array, salp swarm optimization, time moment matching method

## 1. Introduction

In the system engineering and control systems, the order of the plant or designed controller could be very high, which adds a burden on simulation and computation. If a higher-order system or controller can be diminished in terms of its order, it saves computation time and cost. Order diminution (OD) or model order reduction (MOR) may be classified into three headings: classical methods, using optimization algorithms, and mixed methods.

The classical methods are proposed by different researchers such as Davison [1], Marshall [2], Fossard [3], Davidson and Lucas [4], and Krishnamurthy and Seshadri [5]. These techniques are based on some simple concepts such as dominant pole preservation, dividing state matrix into two parts, retention of predominated eigenvalues, continued fraction expansions, and Routh stability criterion. The disadvantages of these methods are the higher cost of computation and sometimes the problem of stability. These classical methods are also mixed by many researchers to get a simplified system [6, 7]. Recently, an improved balanced realization technique

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[8] was proposed in which the denominator was calculated based on the balanced realization, whereas the numerator was determined by the simple mathematical procedure.

In the second method, optimization techniques are used for order reduction. "They are based on minimizing the error between the source system and the reduced system" [9]. The error functions are taken as integral of absolute error (IAE) [10], integral of time multiplied by absolute error (ITAE) [11], integral of square error (ISE) [10], integral of time multiplied by square error (ITSE) [11], L1, L2, L-infinity norm [12], weighted error function [13], etc. Meta-heuristic algorithms inspired by nature are widely used in the optimization technique method [14]. First, in this category, comes a genetic algorithm (GA) [15] that works on the survival of the fittest. Some other models are particle swarm optimization (PSO) [16] based on flocking behavior of birds, harmony search algorithm (HSA) [17] based on musical improvisation, cuckoo search algorithm (CSA) $[18,19]$ based on a cuckoo bird, which lays their own eggs in the nest of other host bird, and Big Bang-Big Crunch [20] theory based on the evolution of the Universe. Generally, these methods do not retain dominant poles, which is a serious drawback.

In the third category, order reduction is performed by mixing classical methods and optimization algorithms. Such a method was proposed by Nadi [21] for single-input single-output (SISO) and multi-input multi-output (MIMO) systems in which dominant poles were chosen from the original system, while the other parameters were selected by PSO. Big Ban-Big Crunch optimization was mixed with Routh approximation by Desai et al. [22] and recently, it was also mixed with time moment matching [20]. In the above two methods, the numerator was calculated by Big Bang-Big Crunch optimization, and the denominator was estimated by Routh approximation and time moment matching, respectively. A unified hybrid metaheuristic algorithm [23] for the discrete-time system was proposed by Ganguli in which a gray wolf optimization and firefly algorithm were combined for order reduction.

The order diminution technique proposed in this article uses the time moment matching method with the Routh array concept and the recently developed fast and accurate salp swarm optimization (SSO) technique [20]. Time moment matching criterion ensures the dominant characteristic of the original system in the corresponding diminished system, and it is a very reliable technique for reducing the complexity of a given problem. This concept has been used by many researchers such as Pal [24], Shamash [6], and Biradar [20]. The denominator coefficients (in the frequency domain) of the diminished system are calculated with the help of the time moment matching method using the Routh array concept, which ensures the stable nature of the original system. On the other hand, the numerator of the diminished system is obtained by utilizing the fast and better search capability of the SSO technique using integral of square error (ISE) as an objective function. The SSO ensures the minimum error between the diminished and the original system. A variety of benchmark problems are considered here to show the superiority of this technique. The first problem is a SISO system of $8^{t h}$ order, and the second one has a time delay. A very high order $\left(\left(84^{t h}\right)\right)$ transfer function is taken as the third example, which is approximated to a $2^{\text {nd }}$ order system using the proposed technique. The obtained results of all problems are intensively compared with other classical as well as optimization methods provided in the literature; the method proposed in this paper provides the least ISE and the best approximated reduced system.

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## 2. Problem background

### 2.1. General model order diminution

Let us define a system transfer function of order $n$ as

$$
\begin{equation*}
G(s): u \rightarrow y \tag{1}
\end{equation*}
$$

The objective of order diminution is to find out a new transfer function $G_{r}(s)$ of diminished order $n_{r}$ which is defined as

$$
\begin{equation*}
G_{r}(s): u \rightarrow y_{r} \text { with } n_{r}<n \tag{2}
\end{equation*}
$$

provided input and output characteristics are the same for the above two systems, that is for the same input $u(t)$, output $y(t) \approx y_{r}(t)$.

### 2.2. Model order diminution in terms of transfer matrix/function for MIMO and SISO systems

Let a higher-order MIMO system of order $n$ is denoted by the following transfer matrix

$$
[G(s)]=\left[\begin{array}{ccc}
G_{11}(s) & \cdots & G_{1 j}(s)  \tag{3}\\
\vdots & \ddots & \vdots \\
G_{i 1}(s) & \cdots & G_{i j}(s)
\end{array}\right] ; \quad i, j>0
$$

where, $i=$ no of inputs, $j=$ no of outputs and $G_{i j}(s)=$ transfer function for $i^{\text {th }}$ input and $j^{t h}$ output. Equation (3) can be expressed in terms of numerator and denominator polynomials as

$$
\begin{align*}
& {[G(s)]=\left[\begin{array}{ccc}
\frac{N_{11}(s)}{D_{11}(s)} & \cdots & \frac{N_{1 j}(s)}{D_{1 j}(s)} \\
\vdots & \ddots & \vdots \\
\frac{N_{i 1}(s)}{D_{i 1}(s)} & \cdots & \frac{N_{i j}(s)}{D_{i j}(s)}
\end{array}\right]}  \tag{4}\\
& \quad=\frac{1}{D(s)}\left[\begin{array}{ccc}
N_{11}(s) & \cdots & N_{1 j}(s) \\
\vdots & \ddots & \vdots \\
N_{i 1}(s) & \cdots & N_{i j}(s)
\end{array}\right] \tag{5}
\end{align*}
$$

where $N_{i j}(s)$ and $D_{i j}(s)$ are the numerator and denominator polynomials corresponding to $i^{t h}$ input and $j^{t h}$ output; $D(s)$ represents a common denominator. The transfer function $G_{i j}(s)$ can be expressed as

$$
\begin{equation*}
G_{i j}(s)=\frac{a_{0} s^{m}+a_{1} s^{m-1}+a_{2} s^{m-2}+\cdots a_{m} s^{0}}{b_{0} s^{n}+b_{1} s^{n-1}+b_{2} s^{n-2}+\cdots b_{n} s^{0}},(m \leq n) \tag{6}
\end{equation*}
$$

If we put $i=1$ and $j=1$ in Equation (6), the same equation represents the transfer function of the SISO system with $G_{i j}(s)=G_{11}(s)$ or $G(s)$.
Then the reduced-order transfer function corresponding to $G_{i j}(s)$ is denoted by the following equation

$$
\begin{equation*}
G_{i j, r}(s)=\frac{c_{0} s^{m_{r}-1}+c_{1} s^{m_{r}-2}+c_{2} s^{m_{r}-3}+\ldots c_{m_{r}-1} s^{0}}{d_{0} s^{n_{r}}+d_{1} s^{n_{r}-1}+d_{2} s^{n_{r}-2}+\ldots d_{n_{r}} s^{0}}, m_{r} \leq n_{r} \tag{7}
\end{equation*}
$$

In the case of the SISO system, $G_{i j r}(s)$ becomes $G_{r}(s)$. The Eqn. (7) may also be transformed into the state space form if required.

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### 2.3. Order diminution approach applied in optimization technique

Let for input $u(t)$, the output of the original full-order system (Equation (6)) be $y(t)$ and for the same input, the output of the reduced system (Equation (7)) be $y_{r}(t)$. Then, the error between the two systems is estimated by the following equation

$$
\begin{equation*}
e(t)=y(t)-y_{r}(t) \tag{8}
\end{equation*}
$$

The reduced-order system is obtained by minimizing the error criterion named as integral of square error [10]

$$
\begin{equation*}
I S E=\int_{0}^{T_{s s}} e(t)^{2} d t \tag{9}
\end{equation*}
$$

where $T_{s s}$ is the steady-state time.
Remark 1: Only Linear time-invariant systems with poles on the left-hand side of s-planes are considered in this article.

## 3. Theoretical background of proposed approach

The denominator of the diminished system is calculated by using the time moment matching method with the Routh array concept, whereas the numerator is obtained with the help of the fast and more accurate salp swarm optimization algorithm. Integral of square error (ISE) between the original and diminished systems is selected as the objective function to get the numerator coefficient. The details of these two methods are given below.

### 3.1. Time moment matching

For time moment matching [20] consider a system with following transfer function

$$
\begin{equation*}
G(s)=\int_{0}^{\infty} y_{\delta}(t) e^{-s t} d t \tag{10}
\end{equation*}
$$

where $y_{\delta}(t)$ represents the impulse response of $G(s)$. By using the power series expansion of $e^{-s t}$, Equation (10) can be written as

$$
\begin{align*}
G(s) & =\int_{0}^{\infty} y_{\delta}(t)\left\{1-s t+\frac{s^{2} t^{2}}{2!}-\frac{s^{3} t^{3}}{3!}+\cdots\right\} d t  \tag{11}\\
& =\int_{0}^{\infty} y_{\delta}(t) d t-s \int_{0}^{\infty} t y_{\delta}(t) d t+s^{2} \int_{0}^{\infty} \frac{t^{2}}{2!} y_{\delta}(t) d t-\cdots
\end{align*}
$$

In general, Equation (11) can be expressed as

$$
\begin{equation*}
G(s)=k_{0}+k_{1} s+k_{2} s^{2}+\cdots \tag{12}
\end{equation*}
$$

where coefficients $k_{i}$

$$
k_{i}=\frac{(-1)^{i}}{i!} \int_{0}^{\infty} t^{i} y_{\delta}(t) d t, \quad i=0,1,2, \ldots
$$

and $\int_{0}^{\infty} t^{i} y_{\delta}(t) d t$ denote the $i^{t h}$ moment of $y_{\delta}(t)$.
Thus, the time moments of the diminished system are proportional to time moments of the original system and

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time moment matching method aims to match as many moments as possible.
Remark 2: Time moment matching criterion ensures the dominant characteristic of the original system in the corresponding diminished system [20].

### 3.2. Salp swarm optimization (SSO)

Salps, transparent, barrel-shaped fishes belong to the family Salpidae. They resemble jellyfishes. The movement of salp is very similar to jellyfish, in which the water is pumped through the body as propulsion to move forward. The swarming behavior is very important for these fishes. In the deep ocean, salps form a swarm known as a salp chain. As per the researchers, this salp chain is formed for achieving better locomotion using rapid coordinated changes and hunting.
The mathematical model of the swarming behavior of these salps is proposed by Mirjalili [25]. According to this model, salps are divided into two categories: leaders and followers. The salp at the front of the salp chain is known as a leader while the rest slaps are known as followers. The position of salps can be represented in the n-dimensions search space, where $n$ is the number of variables in the problem. Therefore, the position of salps can be stored in two dimensions of space. The swarm's target can be assumed as a food source denoted by T. To update the position of the leader, following equation is used

$$
x_{j}^{1}=\left\{\begin{array}{l}
T_{j}+c_{1}\left(\left(U b_{j}-L b_{j}\right) c_{2}+L b_{j}\right) c_{3} \geq 0  \tag{13}\\
T_{j}-c_{1}\left(\left(U b_{j}-L b_{j}\right) c_{2}+L b_{j}\right) c_{3}<0
\end{array}\right.
$$

where $x_{j}^{1}$ is the position of leader salp, $T_{j}$ is the position of the food source, $U b_{j}$ and $L b_{j}$ are the upper and lower bounds in the j -th dimensions respectively, $c_{1}, c_{2}$, and $c_{3}$ are the random numbers. The coefficient $c_{1}$, which is responsible for exploration and exploitation, is given as

$$
\begin{equation*}
c_{1}=2 e^{-\left(\frac{4 l}{L}\right)^{2}} \tag{14}
\end{equation*}
$$

where $l$ and $L$ are the current and total number of iterations, respectively. The coefficients $c_{1}$ and $c_{2}$ are randomly generated numbers in the interval of $[0,1]$. To update the position of followers, the following equation is used

$$
\begin{equation*}
x_{j}^{i}=\frac{1}{2} a t^{2}+v_{0} t ; \quad i \geq 2 \tag{15}
\end{equation*}
$$

where $x_{j}^{i}$ represents the position of i -th follower in j -th dimension, t is the time, $v_{0}$ is the initial velocity, and $a$ is calculated as follows:

$$
\begin{equation*}
a=\frac{v_{\text {final }}}{v_{0}}, \text { where } v=\frac{x-x_{0}}{t} \tag{16}
\end{equation*}
$$

In the optimization problem, time represents iterations; the difference between two iterations is equal to 1 and if the initial velocity $v_{0}$ is taken as zero, the above equation can be modified as

$$
\begin{equation*}
x_{j}^{i}=\frac{1}{2}\left(x_{j}^{i}+x_{j}^{i-1}\right), \quad i \geq 2 \tag{17}
\end{equation*}
$$

## 4. Schema of the proposed approach

### 4.1. Determine coefficients of diminished denominator polynomial ( $d_{0}, d_{1}, d_{2}, \ldots, d_{n_{r}}$ )

Apply time moment matching technique with Routh array concept to find out denominator of diminished system Step 1: Divide equation (6) by $b_{n}$ to get the following equation

$$
\begin{equation*}
G_{i j}(s)=\frac{a_{0} / b_{n} s^{m}+a_{1} / b_{n} s^{m-1}+a_{2} / b_{n} s^{m-2}+\ldots a_{m-2} / b_{n} s^{2}+a_{m-1} / b_{n} s^{1}+a_{m} / b_{n} s^{0}}{b_{0} / b_{n} s^{n}+b_{1} / b_{n} s^{n-1}+b_{2} / b_{n} s^{n-2}+\ldots b_{m-2} / b_{n} s^{2}+b_{n-1} / b_{n} s^{1}+s^{0}} \tag{18}
\end{equation*}
$$

The above equation can be simplified as:

$$
\begin{equation*}
G_{i j}(s)=\frac{q_{21}+q_{22} s^{1}+q_{23} s^{2}+q_{24} s^{3} \ldots+q_{2 m} s^{m}}{1+q_{12} s^{1}+q_{13} s^{2}+q_{14} s^{3} \ldots+q_{1 n} s^{n}} \tag{19}
\end{equation*}
$$

In the similar manner, the transfer function of the reduced system Eqn (7) is written as:

$$
\begin{equation*}
G_{i j, r}(s)=\frac{\widetilde{q}_{21}+\widetilde{q}_{22} s^{1}+\widetilde{q}_{23} s^{2}+\widetilde{q}_{24} s^{3} \ldots+\widetilde{q}_{2 m_{r}} s^{m_{r}}}{1+\widetilde{q}_{12} s^{1}+\widetilde{q}_{13} s^{2}+\widetilde{q}_{14} s^{3} \ldots+\widetilde{q}_{1 n_{r}} s^{n_{r}}} \tag{20}
\end{equation*}
$$

Step 2: Now, using Eqn (19) makes the following Routh array

$$
\begin{array}{|cccccc}
\hline 1 & q_{12} & q_{13} & q_{14} & \cdots & q_{1 n}  \tag{21}\\
q_{21} & q_{22} & q_{23} & q_{24} & \cdots & q_{2 m} \\
q_{31} & q_{32} & q_{33} & q_{34} & \cdots & q_{3 m} \\
q_{41} & q_{42} & q_{43} & q_{44} & \cdots & q_{4 m} \\
q_{51} & q_{52} & q_{53} & q_{54} & \cdots & q_{5 m} \\
\vdots & \vdots & \vdots & \vdots & \ddots & \vdots
\end{array}
$$

in general $q_{k, l}=q_{k-1,1} \times q_{1, l+1}-q_{k-1, l+1} \times q_{1,1} \forall k=3,4,5, \ldots l=1,2,3, \ldots$, and $q_{11}=1$. The first column of Routh array is used to get the following original system

$$
\begin{equation*}
G_{i j}(s)=q_{21}-q_{31} s+q_{41} s^{2}-q_{51} s^{3}+\cdots \tag{22}
\end{equation*}
$$

Step 3: By using $k_{i}=(-1)^{k+2} \times q_{k, 1} \quad \forall k=2,3,4, \ldots$ and $i=0,1,2, \ldots$, , form the following matrix

$$
\left[\begin{array}{c}
k_{0}  \tag{23}\\
k_{1} \\
k_{2} \\
\vdots \\
k_{m_{r}} \\
\hline k_{m_{r}+1} \\
k_{m_{r}+2} \\
\vdots \\
k_{m_{r}+n_{r}}
\end{array}\right]=\left[\begin{array}{cccccc|ccc}
0 & 0 & \ldots & \ldots & 0 & 0 & 0 & \ldots & 0 \\
-k_{0} & 0 & \ldots & \ldots & 0 & 0 & & & \\
-k_{1} & -k_{0} & \ldots & \ldots & 0 & 0 & \vdots & \ddots & \vdots \\
\vdots & \vdots & \ldots & \ldots & \vdots & 0 & & & \\
-k_{m_{r}-1} & -k_{m_{r}-2} & \ldots & \ldots & -k_{0} & 0 & 0 & \ldots & 0 \\
\hline-k_{m_{r}} & -k_{m_{r}-1} & \ldots & -k_{1} & -k_{0} & 0 & 0 & \ldots & 0 \\
-k_{m_{r}+1} & -k_{m_{r}} & \ldots & -k_{2} & -k_{1} & 0 & \vdots & \ddots & \vdots \\
\vdots & \vdots & \vdots & \ldots & \ldots & \vdots & \vdots & & \vdots \\
-k_{m_{r}+n_{r}+1} & -k_{m_{r}+n_{r}} & \ldots & \ldots & \ldots & 0 & 0 & \ldots & 0
\end{array}\right] \times\left[\begin{array}{c}
\widetilde{q}_{12} \\
\widetilde{q}_{13} \\
\widetilde{q}_{14} \\
\vdots \\
\widetilde{q}_{1, n_{r}+1} \\
0 \\
0 \\
\vdots \\
0
\end{array}\right]+\left[\begin{array}{c}
\widetilde{q}_{21} \\
\widetilde{q}_{22} \\
\widetilde{q}_{23} \\
\vdots \\
\widetilde{q}_{2, m_{r}+1} \\
0 \\
0 \\
\vdots \\
0
\end{array}\right]
$$

In compact form Eqn (23) can be written as:

$$
\left[\begin{array}{c|c}
\widetilde{K}_{1}  \tag{24}\\
\hline \widetilde{K}_{2}
\end{array}\right]=\left[\begin{array}{c|c}
K_{11} & K_{12} \\
\hline K_{21} & K_{22}
\end{array}\right] \times\left[\begin{array}{c}
\widetilde{Q}_{1} \\
\hline 0
\end{array}\right]+\left[\begin{array}{c}
\widetilde{Q}_{2} \\
\hline 0
\end{array}\right]
$$

Step 4: Calculate the value of vector $\widetilde{Q}_{1}=K_{21}{ }^{-1} \times \widetilde{K}_{2}=\left[\begin{array}{llll}\widetilde{q}_{12} & \widetilde{q}_{13} & \cdots & \widetilde{q}_{1 n_{r}}\end{array}\right]^{T}$.
Step 5: The denominator of a reduced $2^{n d}$ order system is given by $\widetilde{q}_{12} s^{2}+\widetilde{q}_{13} s+1$; for $3^{r d}$ order $\widetilde{q}_{12} s^{3}+$ $\widetilde{q}_{13} s^{2}+\widetilde{q}_{14} s+1$, etc. Hence, $\left[\widetilde{q}_{12}, \widetilde{q}_{13}, \ldots \widetilde{q}_{1 n_{r}}\right]$ are the denominator coefficients $\left(d_{0}, d_{1}, d_{2}, \ldots, d_{n_{r}}\right)$ of Eqn (7).

### 4.2. Determine coefficients of diminished numerator polynomial $\left(c_{0}, c_{1}, c_{2}, \ldots, c_{m_{r}}\right)$

Step 1: Randomly initialize the salp population.
Step 2: Put the denominator calculated by the time moment method in Eqn (7) and find out the reduced system.
Step 3: Calculate the error, which is the difference between the original and the reduced system outputs for the step input, using the Eqn (8). Evaluate the fitness of each search agent (salp) using objective function as Eqn (9).
Step 4: Select $T=$ salp with the lowest ISE value.
Step 5: For every salp, update the position of leader salp using Eqn (13) and follower salp by using Eqn (17).

Step 6: Amend the salps based on lower and upper bounds of variables.
Step 7: If maximum number of iterations is reached, return the best solution $\mathrm{T},\left(c_{0}, c_{1}, c_{2}, \ldots, c_{m_{r}}\right)$, otherwise, go to step 3 .
Remark 3: ISE is considered as an objective function because it eliminates errors in transient as well as steady-state response quickly as compared to other error criteria.

## 5. Numerical examples, results, and discussion

This section discusses three different examples showing the superiority of the proposed MOR method. The first example is the $8^{\text {th }}$ order SISO system, whereas the second example has a time delay which becomes a nonminimum system after Pade approximation. The third example has the $84^{t h}$ order. The results of all reduced systems are compared chronically with other up-to-date order reduction techniques available in the literature. The parameters taken for the salp swarm optimization technique are shown in Table 1.

Remark 4: Several trials have been done to get the numerator coefficients using SSO and the best solutions are selected. The lower and upper limits for variables are selected using these trials.

Table 1. SSO parameters.

| Parameters | values |
| :--- | :--- |
| Search agents (Number of salps, N) | 30 |
| Maximum iterations count $(\mathrm{T})$ | 1000 |
| Number of variables $(\operatorname{dim})$ | Problem dependent |
| Random number | $[0,1]$ |
| Lower and upper bounds | -10000 and +10000 |

Example 1 Consider the 8 th order problem [26] with a transfer function given by the following equation

$$
\begin{equation*}
G(s)=\frac{35 s^{7}+1086 s^{6}+13285 s^{5}+82402 s^{4}+278376 s^{3}+511812 s^{2}+482964 s+194480}{s^{8}+33 s^{7}+437 s^{6}+3017 s^{5}+11870 s^{4}+27470 s^{3}+37492 s^{2}+28880 s+9600} \tag{25}
\end{equation*}
$$

First of all, apply the time moment method to find out the denominator and then apply the SSO technique for estimating the numerator as discussed in the proposed schema. The obtained $4^{\text {th }}$ order reduced system has the following transfer function:

$$
\begin{equation*}
G_{r}(s)=\frac{4.178 s^{3}+22.48 s^{2}+34.74 s+20.26}{0.1209 s^{4}+0.8606 s^{3}+1.98 s^{2}+2.24 s+1} \tag{26}
\end{equation*}
$$

The step and Bode graphs are shown in Figures 1 and 2 for the original and reduced systems. It is clear that the proposed system coincides with the original system throughout the entire time and frequency domain. Specifications in terms of transient response and the ISE are shown in Table 2 for comparison purposes. The value of ISE is the least $\left(4.2241 \times 10^{-05}\right)$ and rise time, settling time, and overshoot are matched more closely with the original system as compared to other simplified systems. We can conclude that the proposed method depicts the original system more closely and accurately.


Figure 1. Time response of different systems of Example 1.


Figure 2. Frequency response of different systems of Example 1.

Example 2 Consider the 7 th order delayed system with following transfer function [20]

$$
\begin{equation*}
G(s)=\frac{(4000 s+50000) e^{-0.3 s}}{s^{7}+69 s^{6}+1764 s^{5}+20280 s^{4}+102500 s^{3}+221375 s^{2}+187500 s+50000} \tag{27}
\end{equation*}
$$

Table 2. ISE and transient response comparison for different systems of Example 1.

| Systems | ISE | Rise time <br> $(\mathrm{Sec})$ | Settling time <br> $(\mathrm{Sec})$ | \% Overshoot |
| :--- | :--- | :--- | :--- | :--- |
| Original 25 | - | 1.0725 | 1.5824 | $0.6421 \times 10^{+00}$ |
| Proposed (26) | $4.2239 \times 10^{-05}$ | 1.0728 | 1.5831 | $0.6425 \times 10^{+00}$ |
| Prajapati(2019)[8] | $0.0024 \times 10^{-00}$ | 1.0643 | 1.5777 | $0.5909 \times 10^{+00}$ |
| Prajapati(2018)[26] | $0.0674 \times 10^{-00}$ | 1.0473 | 1.5058 | $1.3692 \times 10^{+00}$ |
| Sikander(2017)[29] | $1.6403 \times 10^{+08}$ | 0.0001 | 4.5413 | $1.6403 \times 10^{+04}$ |
| Singh(2016)[30] | $9.9101 \times 10^{+04}$ | 0.0847 | 9.2898 | $1.4997 \times 10^{+01}$ |
| Sikander(2015)[31] | $2.1448 \times 10^{+03}$ | 5.6520 | 1.3938 | $1.4997 \times 10^{+01}$ |
| Kranthi(2013)[32] | $9.7994 \times 10^{+11}$ | 0.0000 | 3.7559 | $4.3290 \times 10^{+04}$ |
| Vishwakarma(2013)[33] | $1.3550 \times 10^{+11}$ | 0.0000 | 5.6070 | $1.2794 \times 10^{+04}$ |
| Vishwakarma(2008)[34] | $3.2523 \times 10^{+11}$ | 0.0000 | 4.9212 | $2.0647 \times 10^{+06}$ |
| Nidhi(2006)[35] | $8.6722 \times 10^{+05}$ | 0.0060 | 8.7509 | $1.3854 \times 10^{+01}$ |
| Sinha(1990)[36] | $4.6388 \times 10^{+01}$ | 0.4506 | 2.7144 | $5.9124 \times 10^{+01}$ |
| Gutman(1982)[37] | $2.6071 \times 10^{-00}$ | 0.9607 | 1.7605 | $0.0000 \times 10^{+00}$ |
| Moore(1981)[38] | $1.0978 \times 10^{-05}$ | 1.0709 | 1.5780 | $0.6944 \times 10^{+00}$ |
| Shamash(1981)[39] | $1.2542 \times 10^{+02}$ | 2.4455 | 8.5958 | $3.1687 \times 10^{+01}$ |
| Chen(1980)[40] | $2.1463 \times 10^{+10}$ | 0.0001 | 10.654 | $5.1208 \times 10^{+05}$ |
| Chen(1979)[41] | $3.0973 \times 10^{-00}$ | 0.9853 | 7.1210 | $5.9752 \times 10^{+00}$ |
| Krishnamurthy(1978)[5] | $1.5067 \times 10^{-00}$ | 6.5973 | 3.2393 | $8.3407 \times 10^{+00}$ |
| Shamash(1975)[42] | $3.6452 \times 10^{+07}$ | 8.5353 | 1.3255 | $2.5623 \times 10^{+04}$ |
| Shamash(1974)[43] | $7.8388 \times 10^{+10}$ | 6.1306 | 1.3255 | $9.8139 \times 10^{+05}$ |

In the above system, the time delay of 0.3 s can be approximated to the 3 rd order by Pade approximation, so system given by Eqn. (27) converts to 10 th order of the form $G(s)=N(s) / D(s)$, where $N(s)$ and $D(s)$ are given below:

$$
\begin{align*}
& N(s)=-4000 s^{4}+110000 s^{3}-666700 s^{2}-15560000 s+222200000 \\
& D(s)=s^{10}+109 s^{9}+5191 s^{8}+141300 s^{7}+2396000 s^{6}+25680000 s^{5}+167500000 s^{4}+610500000 s^{3}  \tag{28}\\
& +1111000000 s^{2}+866700000 s+222200000
\end{align*}
$$

The above $10^{\text {th }}$ order system is diminished to $2^{\text {nd }}$ order, and the simplified transfer function is given as:

$$
\begin{equation*}
G_{r}(s)=\frac{-0.6318 s+1.002}{2.927 s^{2}+3.377 s+1} \tag{29}
\end{equation*}
$$

The step and Bode graphs are shown in Figures 3 and 4, respectively. It can be noted from Bode plot that all methods are not reliable for higher frequencies, whereas they are performing well for a lower range of frequencies. The step response is matching more closely as compared to reduced models given by Biradar and Desai. A comparison, in terms of ISE and transient time specifications, is shown in Table 3, which depicts that the proposed method outperforms in terms of ISE and other parameters.


Figure 3. Time response of different systems of Example 2.


Figure 4. Frequency response of different systems of Example 2.

Table 3. ISE and transient response comparison for different systems of Example 2.

| Systems | ISE | Rise Time(Sec) | Settling Time (Sec) | \% Overshoot |
| :--- | :--- | :--- | :--- | :--- |
| Original 27 | - | 5.3179 | 10.2271 | 0.0000 |
| Proposed (29) | 0.0019 | 5.5231 | 10.1890 | 0.0000 |
| Biradar(2016)[20] | 0.0029 | 5.4573 | 10.3192 | 0.0000 |
| Desai(2013)[22] | 0.0593 | 6.2732 | 10.2156 | 0.0000 |

Example 3 Consider another benchmark problem of 84th order [27] represented by following state-space model.

$$
A=\left[\begin{array}{cccccccccc}
A_{1} & 0 & 0 & 0 & 0 & 0 & 0 & a_{78} & 0 & 0 \\
0 & A_{2} & 0 & 0 & 0 & 0 & 0 & 0 & \ddots & 0 \\
0 & 0 & A_{3} & 0 & 0 & 0 & 0 & 0 & 0 & a_{154} \\
0 & 0 & 0 & A_{4} & 0 & 0 & 0 & 0 & 0 & 0 \\
0 & 0 & 0 & 0 & A_{5} & 0 & 0 & 0 & 0 & 0 \\
0 & 0 & 0 & 0 & 0 & A_{6} & 0 & 0 & 0 & 0 \\
0 & 0 & 0 & 0 & 0 & 0 & A_{7} & 0 & 0 & 0 \\
a_{1} & 0 & 0 & 0 & 0 & 0 & 0 & A_{8} & 0 & 0 \\
0 & \ddots & 0 & 0 & 0 & 0 & 0 & 0 & \ddots & 0 \\
0 & 0 & a_{77} & 0 & 0 & 0 & 0 & 0 & 0 & A_{42}
\end{array}\right]
$$

where,

$$
A_{i}=\left[\begin{array}{lc}
-734 & 171 \\
-9 & -734
\end{array}\right] ; \quad i=1,2 \ldots 42
$$

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and $a_{j}=196 ; \quad j=1,2 \ldots 154$
$B=\left[B_{i, 1}\right]_{84 \times 1} ; \quad i=1,2 \ldots 84$
The values of $\left[B_{i}\right]$ are given in Table 5 in the appendix,

$$
\begin{aligned}
C & =\left[C_{1, j}\right]_{1 \times 84} ; \\
& =B^{T} \\
D & =[0]
\end{aligned} \quad \text { By applying the technique proposed in this paper, the obtained reduced }
$$

system has the following transfer function:

$$
\begin{equation*}
G_{r}(s)=\frac{0.01538 s+8.824}{5.490 \times 10^{-06} s^{2}+5.079 \times 10^{-03} s+1} \tag{30}
\end{equation*}
$$

The same 84th order system can also be reduced by using particle swarm optimization (PSO) resulting in following 2 nd order transfer function

$$
\begin{equation*}
G_{r P S O}(s)=\frac{99.9 s+99.24}{0.03775 s^{2}+11.36 s+11.33} \tag{31}
\end{equation*}
$$

and by genetic algorithm(GA)

$$
\begin{equation*}
G_{r G A}(s)=\frac{22.24 s+7.154}{0.008615 s^{2}+2.523 s+0.8108} \tag{32}
\end{equation*}
$$

This higher-order system is also reduced to 2 nd order by Sikandar[28] with the following transfer function

$$
\begin{equation*}
G_{r \text { Sikandar }}(s)=\frac{41.64 s+1.08 \times 10^{7}}{s^{2}+2522 s+1.224 \times 10^{6}} \tag{33}
\end{equation*}
$$

The step responses of the systems are shown in Figure 5 and Bode graphs in Figure 6. It is evident that the response in time and frequency domains is very well approximated by the proposed technique. This approach provides good approximation at lower as well as at higher frequencies as shown by the Bode plot. For this particular problem of 84 th order, PSO does not perform well $\left(I S E=3.8487 \times 10^{-2}\right)$ as compared to $G A$ $\left(I S E=4.7727 \times 10^{-7}\right)$, and Sikander [28] $\left(I S E=2.1023 \times 10^{-3}\right)$ but proposed technique $\left(I S E=5.2610 \times 10^{-8}\right)$ provides the best results among all as shown in Table 4. This table shows the comparative analysis between the proposed and other optimization methods such as GA and PSO in terms of rise time, settling time, and overshoot. These values are almost equal to the original system by the proposed reduced $2 n d$ order system. It concludes that the proposed technique also works for very high order systems.

## 6. Conclusions

A rigorous study of the new hybrid method for order diminution has been carried out in this article. The technique used here is the conjugation of 2 strategies: time moment matching technique with the concept of the Routh array and ISE error- based salp swarm optimization (SSO) algorithm. The divisor of the reduced system is obtained by using the time moment matching method, whereas the dividend by SSO algorithm taking the integral of time multiplied by square error as the error minimization criterion. The Routh array of time moments matching technique guarantees the stability of the reduced system if the original system is stable and SSO ensures the minimum error between the diminished and the original system. A chronicle comparison of


Figure 5. Time response of different systems of Example 3.


Figure 6. Frequency response of different systems of Example 3.

Table 4. ISE and transient response comparison for different systems of Example 3.

| Systems | ISE | Rise time $(\mathrm{Sec})$ | Settling time $(\mathrm{Sec})$ | \%Overshoot |
| :--- | :--- | :--- | :--- | :--- |
| Original[20] | - | $7.4464 \times 10^{-3}$ | $1.3412 \times 10^{-2}$ | 0 |
| Proposed $(30)$ | $5.2610 \times 10^{-8}$ | $7.4465 \times 10^{-3}$ | $1.3425 \times 10^{-2}$ | 0 |
| PSO $(31)$ | $3.8487 \times 10^{-2}$ | $7.1125 \times 10^{-3}$ | $1.2007 \times 10^{-2}$ | 0 |
| GA $(32)$ | $4.7727 \times 10^{-7}$ | $7.5106 \times 10^{-3}$ | $1.3374 \times 10^{-2}$ | 0 |
| Sikander $[28]$ | $2.1023 \times 10^{-3}$ | $3.6748 \times 10^{-3}$ | $6.6207 \times 10^{-3}$ | 0 |

the proposed technique has been carried out with the variety of benchmark problems ranging from 8 th to $84 t h$ order systems, and the comparative results prove the effectiveness of the proposed methodology. The following conclusions are drawn from this study.

- The range of problems (SISO and time-delayed) solved by the proposed method demonstrates the general nature of the technique.
- The proposed method also works very well on the $84 t h$ order system.
- The resulting $I S E$ error is the lowest with the proposed technique.
- The obtained rise time, settling time, and percentage overshoot of the diminished system are very close to the original system.


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Further, looking into the success of the proposed algorithm, a similar technique may be developed for a nonlinear system. Additionally, a reduced base controller can be designed to reduce the complexity of the design as well as to save simulation costs.

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## Appendix A

Table 5. Coefficients of $\left[B_{i, 1}\right]$

| $i$ | $\left[B_{i, 1}\right]$ | $i$ | $\left[B_{i, 1}\right]$ | $i$ | $\left[B_{i, 1}\right]$ | $i$ | $\left[B_{i, 1}\right]$ | $i$ | $\left[B_{i, 1}\right]$ | $i$ | $\left[B_{i, 1}\right]$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 7.624824 | 15 | 3.413381 | 29 | 3.468259 | 43 | 8.562598 | 57 | 2.056365 | 71 | 9.036562 |
| 2 | 2.354606 | 16 | 5.595597 | 30 | 3.058394 | 44 | 5.995426 | 58 | 2.93394 | 72 | 6.950758 |
| 3 | 7.570281 | 17 | 3.976737 | 31 | 9.01239 | 45 | 4.822466 | 59 | 2.561231 | 73 | 7.567942 |
| 4 | 2.321555 | 18 | 9.456895 | 32 | 1.065961 | 46 | 0.120089 | 60 | 3.126108 | 74 | 2.666435 |
| 5 | 4.608591 | 19 | 8.76504 | 33 | 2.963187 | 47 | 2.787891 | 61 | 5.486768 | 75 | 6.112229 |
| 6 | 0.353796 | 20 | 9.884494 | 34 | 8.515574 | 48 | 5.741295 | 62 | 8.308178 | 76 | 1.722831 |
| 7 | 5.242211 | 21 | 3.913699 | 35 | 7.294136 | 49 | 8.2792 | 63 | 2.974988 | 77 | 1.518785 |
| 8 | 3.992345 | 22 | 9.201152 | 36 | 6.427816 | 50 | 7.045744 | 64 | 8.77164 | 78 | 6.839423 |
| 9 | 8.994307 | 23 | 4.575071 | 37 | 2.681877 | 51 | 3.418584 | 65 | 7.548852 | 79 | 2.179521 |
| 10 | 1.71435 | 24 | 6.616429 | 38 | 7.281299 | 52 | 4.020264 | 66 | 3.25237 | 80 | 7.636844 |
| 11 | 0.247185 | 25 | 5.214498 | 39 | 8.92205 | 53 | 5.389039 | 67 | 2.581167 | 81 | 0.669466 |
| 12 | 5.921826 | 26 | 9.829012 | 40 | 5.099291 | 54 | 1.897445 | 68 | 7.341292 | 82 | 2.401817 |
| 13 | 5.702193 | 27 | 8.996692 | 41 | 5.530897 | 55 | 4.346914 | 69 | 4.374441 | 83 | 3.542152 |
| 14 | 5.824618 | 28 | 6.302044 | 42 | 8.808669 | 56 | 4.086833 | 70 | 3.124567 | 84 | 6.076599 |


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